

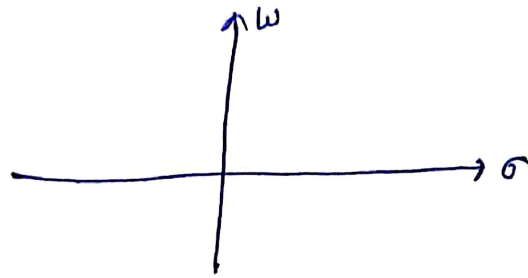
## Laplace Transform

Laplace Transform is a mathematical tool which is used to convert a time domain signal into s-domain.

$$\text{Where } s = \sigma + j\omega$$

$\sigma$  is a real part of  $s$

and  $\omega$  is a imaginary part of  $s$



Laplace Transform is used for continuous time signal and it is also applicable for those signal whose Fourier Transform is not exist.

Laplace Transform comes in two varieties:-

- (i) Bilateral
- (ii) Unilateral

The bilateral Laplace Transform offers insight into nature of system charac. such as stability, causality and Frequency Response.

The Unilateral LT is a convenient tool for solving differential equations with initial conditions.

## The Bilateral Laplace Transform

$$H(s) = \int_{-\infty}^{\infty} h(t) e^{-st} dt$$

$e^{st}$  is called eigenfunction of LTI system  
and  $H(s)$  is called eigenvalues of LTI system.

Now Inverse Laplace

$$h(t) = \frac{1}{2\pi j} \int_{\sigma-j\infty}^{\sigma+j\infty} H(s) e^{st} ds$$

## Region of Convergence (ROC) for Laplace Transform:-

Laplace Transform is guaranteed to converge  
if  $x(t) e^{-\sigma t}$  is absolutely integrable.

$$\int_{-\infty}^{\infty} |x(t) e^{-\sigma t}| dt < \infty$$

The Range of  $R\{s\} = \sigma$  for which the Laplace Transform converges is termed the Region of Convergence.

Q. Find the Laplace Transform of

$$x(t) = e^{-at} u(t)$$

Sol:

$$X(s) = \int_{-\infty}^{\infty} x(t) e^{-st} dt$$
$$= \int_{-\infty}^{\infty} e^{-at} u(t) e^{-st} dt$$

$$\text{Now } u(t) = \begin{cases} 1 & t > 0 \\ 0 & t < 0 \end{cases}$$

$$X(s) = \int_0^{\infty} e^{-at} e^{-st} dt$$
$$= \int_0^{\infty} e^{-t(s+a)} dt$$
$$= \left[ \frac{e^{-t(s+a)}}{-(s+a)} \right]_0^{\infty}$$
$$= -\frac{1}{s+a} [0 - 1] = \frac{1}{s+a}$$

Now Roc:-  $s+a > 0$

$$s > -a$$

Now  $s = \sigma + j\omega$

So  $\sigma + j\omega > -a$

or  $\text{Re}\{s\} > -a$

So  $L[e^{-at} u(t)] = \frac{1}{s+a} \quad \text{Re}\{s\} > -a$

Q. Find Laplace Transform of

$$x(t) = -e^{-at} u(-t)$$

Sol:

$$X(s) = \int_{-\infty}^{\infty} x(t) e^{-st} dt$$
$$= - \int_{-\infty}^{\infty} e^{-at} u(-t) e^{-st} dt$$

$$u(-t) = \begin{cases} 1 & t \leq 0 \\ 0 & t > 0 \end{cases}$$

$$\text{So } X(s) = - \int_{-\infty}^0 e^{-at} e^{-st} dt + 0$$

$$= - \int_{-\infty}^0 e^{t(-s-a)} dt$$

$$= - \left[ \frac{e^{t(-s-a)}}{-s-a} \right]_{-\infty}^0$$

$$= \frac{1}{s+a} [1 - 0] \quad -s-a > 0$$

$$= \frac{1}{s+a} \quad s < -a$$

$$\text{So } X(s) = \frac{1}{s+a} \quad \text{Re}\{s\} < -a$$

Q. Find Laplace of

$$x(t) = e^{-|t|}$$

Sol:

$$|t| = \begin{cases} t & t \geq 0 \\ -t & t < 0 \end{cases}$$

$$\text{So } x(t) = \begin{cases} e^{-t} & t \geq 0 \\ e^t & t < 0 \end{cases}$$

$$\begin{aligned} X(s) &= \int_{-\infty}^0 e^t e^{-st} dt + \int_0^{\infty} e^{-t} e^{-st} dt \\ &= \int_{-\infty}^0 e^{t(1-s)} dt + \int_0^{\infty} e^{-t(1+s)} dt \\ &= \left( \frac{e^{t(1-s)}}{1-s} \right)_{-\infty}^0 + \left( \frac{e^{-t(1+s)}}{-(1+s)} \right)_{0}^{\infty} \end{aligned}$$

$$\Rightarrow \frac{1}{1-s} [1-0] - \frac{1}{1+s} [0-1]$$

$$\Rightarrow \frac{1}{1-s} + \frac{1}{1+s} = \frac{2}{1-s^2}$$

$$\text{Roc: } \begin{aligned} 1-s > 0 & \quad \text{and} \quad 1+s > 0 \\ s < 1 & \quad \text{and} \quad s > -1 \end{aligned}$$

$$\text{So } -1 < s < 1$$

$$\text{or } \boxed{-1 < \text{Re}\{s\} < 1}$$

Q. Find Laplace of

(a)  $x(t) = \sin(\omega_0 t) u(t)$

(b)  $x(t) = \cos(\omega_0 t) u(t)$

Sol: We know that

$$e^{i\omega_0 t} = \cos\omega_0 t + i \sin\omega_0 t$$

So  $\cos\omega_0 t \Rightarrow$  Real part of  $e^{i\omega_0 t}$   
And  $\sin\omega_0 t \Rightarrow$  Imag. part of  $e^{i\omega_0 t}$

Now again we know that

$$L[e^{-at} u(t)] = \frac{1}{s+a} \quad \text{Re}\{s\} > -a$$

$$\text{So } L[e^{i\omega_0 t} u(t)] = \frac{1}{s - j\omega_0} \quad \text{Re}\{s\} > 0$$

$$\Rightarrow L[\cos(\omega_0 t) u(t) + j \sin(\omega_0 t) u(t)] = \frac{s + j\omega_0}{s^2 + \omega_0^2}$$

$$L[\cos(\omega_0 t) u(t)] = \frac{s}{s^2 + \omega_0^2} \quad \text{Re}\{s\} > 0$$

$$\text{And } L[\sin(\omega_0 t) u(t)] = \frac{\omega_0}{s^2 + \omega_0^2} \quad \text{Re}\{s\} > 0$$

Q. Find Laplace Transform of

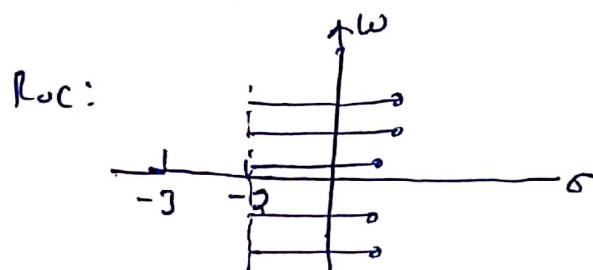
$$x(t) = e^{-2t} u(t) - e^{-3t} u(t)$$

Sol:  $L\{x(t)\} = L\{e^{-2t} u(t) - e^{-3t} u(t)\}$

$$= \frac{1}{s+2} - \frac{1}{s+3}$$

$$\operatorname{Re}\{s\} > -2 \quad \operatorname{Re}\{s\} > -3$$

So  $X(s) = \frac{s+3-s-2}{(s+2)(s+3)} = \frac{1}{s^2+5s+6}$



(Intersection of both)

So  $X(s) = \frac{1}{s^2+5s+6} \quad \operatorname{Re}\{s\} > -2$

Q. Determine Laplace Transform of

$$x(t) = \begin{cases} e^{-at} & 0 < t < T \\ 0 & \text{otherwise} \end{cases}$$

Sol:

$$\begin{aligned} L\{x(t)\} &= \int_{-\infty}^{\infty} x(t) e^{-st} dt \\ &\Rightarrow \int_{-\infty}^0 x(t) e^{-st} dt + \int_0^T x(t) e^{-st} dt + \int_T^{\infty} x(t) e^{-st} dt \\ &\Rightarrow 0 + \int_0^T e^{-at} e^{-st} dt + 0 \\ &\Rightarrow \int_0^T e^{-t(s+a)} dt \end{aligned}$$



$$= \left[ \frac{e^{-t(s+a)}}{-(s+a)} \right]_0^T$$

$$= -\frac{1}{s+a} \left[ e^{-T(s+a)} - 1 \right]$$

$$\boxed{X(s) = \frac{1}{s+a} \left[ 1 - e^{-(s+a)T} \right]}$$

Ans.

The given signal is a Finite duration signal. So Roc is the entire s-plane.

Now at  $s = -a$ ,  $X(s) = \frac{0}{0}$  which is Undefined.

So by L'Hospital Rule

$$\lim_{s \rightarrow -a} X(s) = \lim_{s \rightarrow -a} \left[ \frac{0 + T e^{-(s+a)T}}{1} \right]$$

$$\boxed{X(-a) = T} \quad \text{Ans}$$

$$X(s) = \frac{1}{s+a} \left[ 1 - e^{-(s+a)T} \right] \quad \text{Roc: entire s-plane}$$

T at  $s = -a$  Roc: entire s-plane.



Q Find inverse Laplace of

$$X(s) = \frac{-3}{(s+2)(s-1)}$$

Roc: (a)  $\operatorname{Re}\{s\} > 1$

(b)  $\operatorname{Re}\{s\} < -2$

(c)  $-2 < \operatorname{Re}\{s\} < 1$

Sol: We know that

$$\frac{1}{s+a} \rightarrow e^{-at} u(t) \quad \dots (i)$$

$$\operatorname{Re}\{s\} > -a$$

and  $\frac{1}{s+a} \operatorname{Re}\{s\} < -a \rightarrow -e^{-at} u(-t) \quad \dots (ii)$

We use these results to find out ILT of  $X(s)$

Now given  $X(s) = \frac{-3}{(s+2)(s-1)} = \frac{A}{s+2} + \frac{B}{s-1}$

$$-3 = A(s-1) + B(s+2)$$

$s=1 \quad -3 = B(3)$

$$B = -1$$

And at  $s = -2 \quad -3 = A(-3)$

$$A = 1$$

So  $X(s) = \frac{1}{s+2} - \frac{1}{s-1}$

(a)  $\operatorname{Re}\{s\} > 1$

from formula (i)

$$x(t) = e^{-2t} u(t) - e^t u(t)$$

(b)  $\text{Re}\{s\} < -2$

use formula (ii)

$$\boxed{x(t) = -e^{-2t} u(-t) + e^t u(-t)} \quad \text{Ans.}$$

(c)  $-2 < \text{Re}\{s\} < 1$

for greater than use formula (i) and for less than use formula (ii)

$$\boxed{x(t) = e^{-2t} u(t) + e^t u(-t)} \quad \text{Ans.}$$

### Properties of Laplace Transform:-

(i) linearity:-

$$\text{if } x_1(t) \longleftrightarrow X_1(s) \quad \text{Roc: } R_1$$

$$x_2(t) \longleftrightarrow X_2(s) \quad \text{Roc: } R_2$$

$$\text{Then } ax_1(t) + bx_2(t) \longrightarrow aX_1(s) + bX_2(s) \quad R_1 \cap R_2$$

(ii) Time shifting:-

$$\text{if } x(t) \longleftrightarrow X(s) \quad \text{Roc: } R$$

$$\text{Then } x(t-t_0) \longleftrightarrow X(s) e^{-st_0} \quad \text{Roc: } R$$

(iii) Shifting in s-domain:-

$$\text{if } x(t) \longleftrightarrow X(s) \quad \text{Roc: } R$$

$$\text{Then } x(t) e^{s_0 t} \longrightarrow X(s-s_0) \quad \text{Roc: } R + \text{Re}\{s_0\}$$

Q. Find Laplace Transform of

$$(a) x_1(t) = e^{-at} \cos(\omega_0 t) u(t)$$

$$(b) x_2(t) = e^{-at} \sin(\omega_0 t) u(t)$$

Sol: (a)  $x_1(t) = e^{-at} \cos(\omega_0 t) u(t)$

We know That

$$L[\cos(\omega_0 t) u(t)] = \frac{s}{s^2 + \omega_0^2} \quad \text{Re}\{s\} > 0$$

So, by shifting property

$$L[e^{-at} \cos(\omega_0 t) u(t)] = \frac{s+a}{(s+a)^2 + \omega_0^2} \quad \text{Re}\{s\} > 0 - a$$

So  $L[e^{-at} \cos(\omega_0 t) u(t)] = \frac{s+a}{(s+a)^2 + \omega_0^2} \quad \text{Re}\{s\} > -a$

(b)  $x_2(t) = e^{-at} \sin(\omega_0 t) u(t)$

We know that

$$L[\sin(\omega_0 t) u(t)] = \frac{\omega_0}{s^2 + \omega_0^2} \quad \text{Re}\{s\} > 0$$

Now by s-domain shifting property.

$$L[e^{-at} \sin(\omega_0 t) u(t)] = \frac{\omega_0}{(s+a)^2 + \omega_0^2} \quad \text{Re}\{s\} > -a$$

#### (4) Time scaling Property:-

$$\text{if } x(t) \longleftrightarrow X(s) \quad \text{ROC: } R$$

$$\text{Then } x(at) \longleftrightarrow \frac{1}{|a|} X\left(\frac{s}{a}\right) \quad \text{ROC: } aR$$

Proof We know That

$$X(s) = \int_{-\infty}^{\infty} x(t) e^{-st} dt$$

Case 1: For a positive real constant 'a'

$$L[x(at)] = \int_{-\infty}^{\infty} x(at) e^{-st} dt$$

$$\text{let } at = \tau, \quad dt = \frac{d\tau}{a}$$

$$= \int_{-\infty}^{\infty} x(\tau) e^{-s/a \tau} \frac{d\tau}{a}$$

$$= \frac{1}{a} \int_{-\infty}^{\infty} x(\tau) e^{-\frac{s}{a} \tau} d\tau$$

$$= \frac{1}{a} X(s/a)$$

Case 2:- if a is negative real value.

$$L[x(-at)] = \int_{-\infty}^{\infty} x(-at) e^{-st} dt$$

$$\text{let } -at = \tau$$

$$-a dt = d\tau$$

$$dt = -\frac{1}{a} d\tau$$

$$\Rightarrow \int_{\infty}^{-\infty} x(\tau) e^{s/a \tau} \left(-\frac{d\tau}{a}\right)$$

$$= -\frac{1}{a} \int_{\infty}^{-\infty} x(\tau) e^{-(-s/a)\tau} d\tau$$

$$= \frac{1}{a} \int_{-\infty}^{\infty} x(\tau) e^{-(-s/a)\tau} d\tau$$

$$= \frac{1}{a} X(-s/a)$$

Combining two

$$L[x(at)] = \frac{1}{|a|} x\left(\frac{s}{a}\right)$$

Q. Determine Laplace Transform of

(a)  $g_1(t) = \delta(2t-3)$

(b)  $g_2(t) = u(2t-1)$

(c)  $g_3(t) = \gamma\left(\frac{t}{3}-2\right)$

Sol: (a)  $g_1(t) = \delta(2t-3)$

First we find out LT of  $\delta(t)$

$$L[\delta(t)] = \int_{-\infty}^{\infty} \delta(t) e^{-st} dt$$

$$= \delta(0) = 1$$

so  $L[\delta(t)] = 1$       ROC: entire s plane

Now by time shifting property

$$\delta(t-3) \rightarrow e^{-3s} (1) = e^{-3s}$$

Now again by time scaling property

$$\delta(2t-3) \rightarrow \frac{1}{2} e^{-3s} \quad \text{ROC: entire s plane.}$$

Ans.

$$(b) \quad g_2(t) = u(2t-1)$$

Sol:  $u(t) \xrightarrow{\quad} \frac{1}{s} \quad \text{Re}\{s\} > 0$

Now by shifting

$$u(t-1) \rightarrow e^{-s} \frac{1}{s} \quad \text{Re}\{s\} > 0$$

by scaling

$$u(2t-1) \rightarrow \frac{1}{2} e^{-s/2} \frac{1}{(s/2)}$$

$$= \frac{1}{s} e^{-s/2} \quad \text{Re}\{s\} > 0 \quad \text{Ans.}$$

$$(c) \quad g_3(t) = \gamma \left[ \frac{t}{3} - 2 \right]$$

Sol:  $\gamma(t) = t u(t)$

$$\gamma(t) \xrightarrow{s} \frac{1}{s^2} \quad \text{Re}\{s\} > 0$$

Now by shifting

$$\gamma(t-2) \rightarrow \frac{1}{s^2} e^{-2s} \quad \text{Re}\{s\} > 0$$

Now again by time scaling

$$\gamma(t/3-2) \rightarrow 3 \frac{1}{(3s)^2} e^{-6s} \quad \text{Re}\{s\} > 0$$

$$= \frac{1}{3s^2} e^{-6s} \quad \text{Re}\{s\} > 0$$

Ans.



### ⑤ Time Reversal:-

$$\text{if } x(t) \longleftrightarrow X(s) \quad \text{Roc: } R$$

$$\text{Then } x(-t) \longleftrightarrow X(-s) \quad \text{Roc: } -R$$

### ⑥ Differentiation in the time domain

$$\text{if } x(t) \longleftrightarrow X(s) \quad \text{Roc: } R$$

$$\text{Then } \frac{dx(t)}{dt} \longleftrightarrow sX(s) \quad \text{Roc: } R$$

Proof

$$x(t) = \frac{1}{2\pi j} \int_{\sigma-j\infty}^{\sigma+j\infty} X(s) e^{st} ds$$

differentiate both side w.r.t  $t$

$$\frac{dx(t)}{dt} = \frac{1}{2\pi j} \int_{\sigma-j\infty}^{\sigma+j\infty} X(s) (s e^{st}) ds$$

$$\Rightarrow \frac{dx(t)}{dt} = \frac{1}{2\pi j} \int_{\sigma-j\infty}^{\sigma+j\infty} (sX(s)) e^{st} ds$$

$$\text{So } \frac{dx(t)}{dt} = \mathcal{L}^{-1} [sX(s)]$$

$$\text{or } \mathcal{L} \left[ \frac{dx(t)}{dt} \right] = sX(s)$$

### ⑦ Differentiation in the s-domain:-

$$\text{if } x(t) \longrightarrow X(s) \quad \text{Roc: } R$$

$$\text{Then } t x(t) \longrightarrow -\frac{dX(s)}{ds} \quad \text{Roc: } R$$



Proof We know that

$$X(s) = \int_{-\infty}^{\infty} x(t) e^{-st} dt$$

differentiation both side w.r.t  $s$

$$\frac{dX(s)}{ds} = \int_{-\infty}^{\infty} x(t) (-t e^{-st}) dt$$

$$\frac{dX(s)}{ds} = \int_{-\infty}^{\infty} (-tx(t)) e^{-st} dt$$

$$\text{So } -tx(t) \xrightarrow{L} \frac{dX(s)}{ds}$$

$$\text{Or } tx(t) \xrightarrow{L} -\frac{dX(s)}{ds}$$

Q. Find Laplace Transform of

$$x(t) = t e^{-at} u(t)$$

Sol: We know That

$$L[e^{-at} u(t)] = \frac{1}{s+a} \quad \text{Re}\{s\} > -a$$

Now by diff. Property

$$L[t e^{-at} u(t)] = -\frac{d}{ds} \left[ \frac{1}{s+a} \right]$$

$$= \frac{1}{(s+a)^2} \quad \text{Re}\{s\} > -a$$

## 8 Convolution Property:-

$$\text{if } x_1(t) \xrightarrow{L} X_1(s) \quad \text{ROC} = R_1$$

$$\text{and } x_2(t) \xrightarrow{L} X_2(s) \quad \text{ROC} = R_2$$

$$\text{Then } x_1(t) * x_2(t) \xrightarrow{L} X_1(s) X_2(s) \quad \text{ROC} = R_1 \cap R_2$$

proof

We know that  $\hookrightarrow$

$$x_1(t) * x_2(t) = \int_{-\infty}^{\infty} x_1(\tau) x_2(t-\tau) d\tau$$

take laplace of both side

$$L[x_1(t) * x_2(t)] = \int_{-\infty}^{\infty} \left\{ \int_{-\infty}^{\infty} x_1(\tau) x_2(t-\tau) d\tau \right\} e^{-st} dt$$

$$= \int_{-\infty}^{\infty} \left\{ \int_{-\infty}^{\infty} x_2(t-\tau) e^{-st} dt \right\} x_1(\tau) d\tau$$

$$= \int_{-\infty}^{\infty} e^{-s\tau} x_2(s) x_1(\tau) d\tau \quad \left[ \text{by using time shifting} \right]$$

$$= x_2(s) \int_{-\infty}^{\infty} x_1(\tau) e^{-s\tau} d\tau$$

$$= X_2(s) X_1(s) \quad \text{Ans.}$$

9. Multiplication Property:-

$$\text{if } x_1(t) \xleftrightarrow{L} X_1(s) \quad \text{Roc} = R_1$$
$$\text{And } x_2(t) \xleftrightarrow{L} X_2(s) \quad \text{Roc} = R_2$$

Then

$$x_1(t)x_2(t) \xleftrightarrow{L} \frac{1}{2\pi j} [X_1(s) * X_2(s)]$$

10. Integration in the time domain:-

$$\text{if } x(t) \xleftrightarrow{L} X(s) \quad \text{if Roc} = R$$

$$\text{Then } \int_{-\infty}^t x(\tau) d\tau \xleftrightarrow{L} \frac{X(s)}{s} \quad \text{Roc} \Rightarrow R \cap \{ \text{Re}(s) > 0 \}$$

11. Conjugation Property:-

$$\text{if } x(t) \leftrightarrow X(s)$$

$$\text{Then } x^*(t) \leftrightarrow X^*(s^*)$$

### Causality:-

For a causal LTI system, the impulse response  $h(t) = 0$  for  $t < 0$ . It means signal is right sided.

So for a system with a rational system f/n, causality of the system is equivalent to the ROC being the right half plane to the right of the rightmost pole.

Stability:- An LTI system is stable if and only if the ROC of its system f/n  $H(s)$  includes the  $j\omega$ -axis.

### Stable and Causal System

A causal system with rational system f/n  $H(s)$  is stable if and only if all the poles of  $H(s)$  lie in the left-half of  $s$ -plane.

Q. Find the Inverse Laplace of

$$H(s) = \frac{1}{s^2 - s - 6}$$

(a) if system is causal and unstable

(b) if system is stable and non-causal.

Sol:

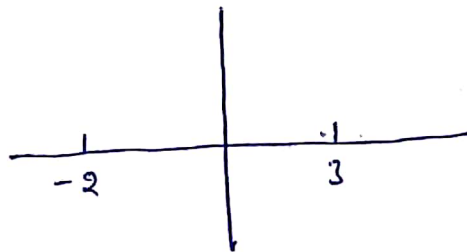
$$H(s) = \frac{1}{s^2 - 3s + 2s - 6} = \frac{1}{s(s-3) + 2(s-3)}$$
$$= \frac{1}{(s+2)(s-3)}$$

$$H(s) = \frac{1}{(s+2)(s-3)} = \frac{A}{s+2} + \frac{B}{s-3}$$

$$A = -\frac{1}{5}$$

$$B = \frac{1}{5}$$

$$\text{So } H(s) = -\frac{1}{5} \frac{1}{s+2} + \frac{1}{5} \frac{1}{s-3}$$



(a) System is causal and unstable  
 Then  $\text{Re}\{s\} > 3$

$$\text{So } h(t) = -\frac{1}{5} e^{-2t} u(t) + \frac{1}{5} e^{3t} u(t) \quad \text{Ans.}$$

(b) System is stable and non-causal  
 Then  $-2 < \text{Re}\{s\} < 3$

$$\text{So } h(t) = -\frac{1}{5} e^{-2t} u(t) - \frac{1}{5} e^{3t} u(-t) \quad \text{Ans.}$$